OTA STATE VARIABLE

OTA_STATE_VARIABLE * dsauersanjose@aol.com 10/21/08			
* www.idea2ic.com *			
*			
* R0 10K * /\ /\ /\			
* \/ \/ IABC= 1mA->0 * R1 1K			
*			
* \/ \/ * \\\ INPO BP R5 20K			
* \OUT0 \ \/ \/ INP1 * \ \ \ \ \ UT1 \LP			
* $//$ $			
$\begin{array}{c ccccccccccccccccccccccccccccccccccc$			
* 7/7 R2 20K 800PF			
* _/\/\/ 800PF			
* INN0 20K R3 * // // //			
*			
$\begin{array}{c c} * & 1 \mathbb{K} & \mathbb{R}4 \\ * & / / / / / / \end{pmatrix}$			
* *	<u> </u>		
*			
VIN	VIN	0 SIN(0 10m 10000) AC 10m
R0	VIN	INPO 10K	· ,
R1 B_OTA0	INP0 OUT0	0 1k 0 I=	-1*v(VIABC)*tanh((v(INP0) -v(INN0))/.052)
CO B BUF0	OUTO BP	0 800p 0 V =	v(OUT0)
$R\overline{2}$	INN0	BP 20k	((()))
R3 R4	INNO INNO	LP 20k 0 1k	
R5	BP	INP1 20k	
R6 B_OTA1	INP1 OUT1	0 1k 0 I=	-1*v(VIABC)*tanh((v(INP1))/.052)
CI B BUF1	OUT1 LP	0 800p 0 V =	V(OUT1)
V_Iabc	VIABC	0 PWL	(0 1m 1m 0)
.control			
tran	1u	1m 0	1u
set plot	pensize = v(vin)	2 v(bp)	v(lp)
op	dec	10 1	1000K
plot	db(bp)	db(lp)	title Gain2OUT@IABC1000uA
alter ac	V_Iabc dec	dc = 10 1	100u 1000K
plot	db(bp)	db(lp)	title Gain2OUT@IABC100uA
alter ac	V_Iabc dec	dc = 10 1	lu 1000K
plot .endc	db(bp)	db(lp)	title Gain2OUT@IABC1uA
.end			

To Covert PDF to plain text click below http://www.fileformat.info/convert/doc/pdf2txt.htm This code works with winspice.

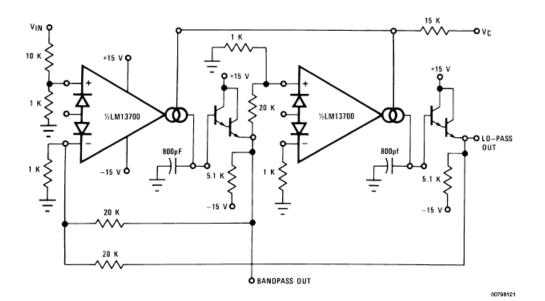
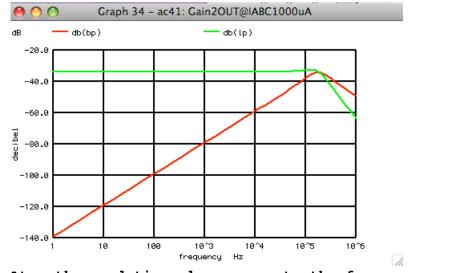


FIGURE 14. Voltage Controlled State Variable Filter

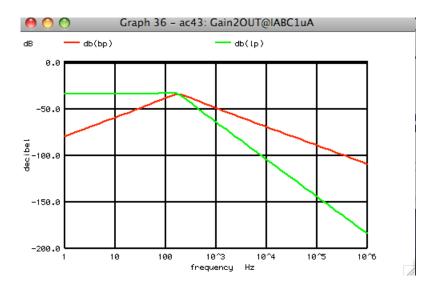
The state variable filters have the beauty to them that the feedback independently defines the shape of the filter. The following two sites show the feedback level differences between Bessel and Butterworth filters.

http://www.idea2ic.com/PlayWithSpice/pdf/Bessel_6P_State_Variable_txt.pdf
http://www.idea2ic.com/PlayWithSpice/pdf/Butterworth 6P State Variable txt.pdf

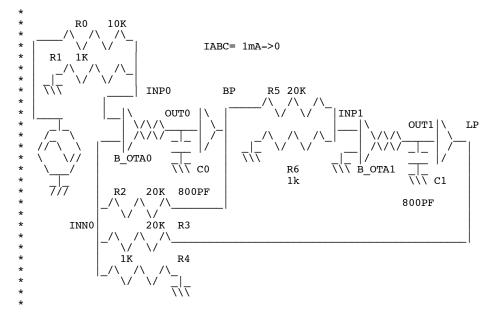
Independently, the gm and C stages define the frequency response. This is seen when the value of IABC is adjusted. At 1mA the response is..



At a thousand times less current, the frequency response is a thousand times lower.



Assuming the OTAs and capacitors and buffers form perfect integrators, there are two different important feedbacks. The feedback from R2 has 90 degrees of phase shift while R3's feedback has 180 degrees of phase shift.

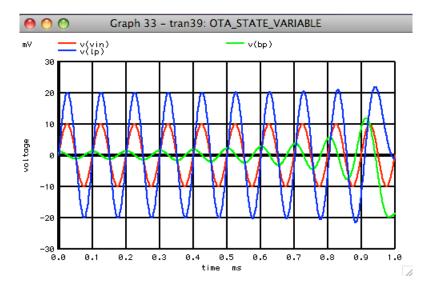


The ratio of 90 degree to 180 degree phase shift feed back obviously sets the stability of the filter. Think of the value of R2 as setting the Q of the filter. The following web site shows that in more detail.

http://www.idea2ic.com/PlayWithSpice/pdf/State Variable s eq 1 txt.pdf

Whenever the integrators of the filter change, the frequency response scales by the same amount.

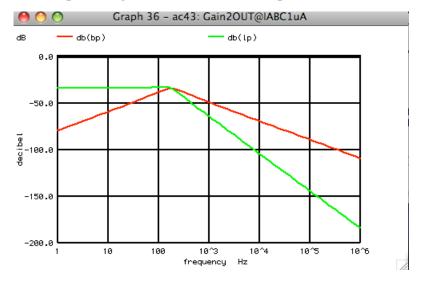
http://www.idea2ic.com/PlayWithSpice/pdf/State_Variable_f_1KHz_txt.pdf



For filters using OTAs, this corresponds to gm an C terms. The above curves show IABC going from 1mA to zero. More information is given below.

http://www.idea2ic.com/PlayWithSpice/pdf/State_Variable_OTA_1KHz_txt.pdf

All state variable filters three outputs. The high pass not shown is just the input signal minus the lowpass.



This filter is using 800pF capacitors. Running at 1uA, the filter is center at a few hundred hertz. A BiCMOS process can easily run down at the 1nA level. And the buffers can use CMOS inputs that have an input impedance of glass. That means the same filter could be built using capacitors 1000 times smaller. That means audio frequency filters can be put completely on chip.